- 1 ELECTRONIC WATERMARKING METHOD AND APPARATUS FOR COMPRESSED
- 2 AUDIO DATA, AND SYSTEM THEREFOR
- 3 Field of the Invention
- 4 The present invention relates to a method and a system for
- 5 embedding, detecting and updating additional information,
- 6 such as copyright information, relative to compressed
- 7 digital audio data, and relates in particular to a technique
- 8 whereby an operation equivalent to an electronic
- 9 watermarking technique performed in a frequency domain can
- 10 be applied for compressed audio data.
- 11 Background Art
- 12 As a technique for the electronic watermarking of audio
- 13 data, there is a Spread Spectrum method, a method for
- 14 employing a polyphase filter, or a method for transforming
- 15 data in a frequency domain and for embedding the resultant
- 16 data. The method for embedding and detecting information in
- 17 the frequency domain has merit in that an auditory
- 18 psychological model can be easily employed, in that high
- 19 tone quality can be easily provided and in that the
- 20 resistance to transformation and noise is high. However,
- 21 the target for the conventional audio electronic
- 22 watermarking technique is limited to digital audio data that
- 23 is not compressed. For the Internet distribution of audio

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- 1 data, generally the audio data are compressed, because of
- 2 the limitation imposed by the communication capacity, and
- 3 the compressed data are transmitted to users. Thus, when
- 4 the conventional electronic watermarking technique is
- 5 employed, it is necessary for the compressed audio data be
- 6 decompressed, for the obtained data to be embedded and for
- 7 the resultant data to be compressed again. The calculation
- 8 time required for this series of operations is extended for
- 9 the advanced audio compression technique that implements
- 10 both high tone quality and high compression efficiency. How
- 11 long it takes before a user can listen to audio data greatly
- 12 effects the purchase intent of a user. Therefore, there is
- 13 a demand for a process whereby the embedding, changing or
- 14 updating of additional information can be performed while
- 15 the audio data are compressed. However, there is presently
- 16 no known method available for embedding additional
- 17 information directly into compressed digital audio data, and
- 18 for changing or detecting the additional information.

19 SUMMARY OF THE INVENTION

- 20 To resolve the above shortcoming, it is one object of the
- 21 present invention to provide a method and a system with
- 22 which information embedded in compressed digital audio data
- 23 can be directly operated.
- 24 It is one more object of the present invention to provide a
- 25 method and a system with which additional information can be

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- 1 embedded in compressed digital audio data.
- 2 It is another object of the present invention to provide a
- 3 method and a system for which only a small memory capacity
- 4 is required in order to embed additional information in
- 5 digital audio data.
- 6 It is an additional object of the present invention to
- 7 provide a method and a system with which minimized
- 8 additional information can be embedded in digital audio
- 9 data.
- 10 It is a further object of the present invention to provide a
- 11 method and a system with which additional information
- 12 embedded in compressed digital audio data can be detected
- 13 without the decompression of the audio data being required.
- 14 It is yet one more object of the present invention to
- 15 provide a method and a system with which additional
- 16 information embedded in compressed digital audio data can be
- 17 changed without the decompression of the audio data being
- 18 required.
- 19 BRIEF DESCRIPTION OF THE DRAWINGS:
- 20 These and other aspects, features, and advantages of the
- 21 present invention will become apparent upon further
- 22 consideration of the following detailed description of the

- 1 invention when read in conjunction with the following
- 2 drawing.
- 3 Fig. 1 is a block diagram illustrating an apparatus for
- 4 embedding additional information directly in compressed
- 5 audio data.
- 6 Fig. 2 is a diagram showing an example for a window length
- 7 and a window function.
- 8 Fig. 3 is a diagram showing the relationship existing
- 9 between a window function and MDCT coefficients.
- 10 Fig. 4 is a block diagram of an MDCT domain that corresponds
- 11 to a frame along a time axis.
- 12 Fig. 5 is a specific diagram showing a sine wave.
- 13 Fig. 6 is a diagram showing an example for embedding
- 14 additional information in an adjacent frame.
- 15 Fig. 7 is a diagram showing a portion of a basis for which
- 16 the MDCT has been performed.
- 17 Fig. 8 is a diagram showing an example of the separation of
- 18 a basis.
- 19 Fig. 9 is a block diagram showing an additional information
- 20 embedding system according to the present invention.

- 1 Fig. 10 is a block diagram showing an additional information
- 2 detection system according to the present invention.
- 3 Fig. 11 is a block diagram showing an additional information
- 4 updating system according to the present invention.
- 5 Fig. 12 is a diagram showing the general hardware
- 6 arrangement of a computer.
- 7 Description of the Symbols
- 8 1: CPU
- 9 2: Bus

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- 10 4: Main memory
- 11 5: Keyboard/mouse controller
- 12 6: Keyboard
- 13 7: Pointing device
- 14 8: Display adaptor card
- 15 9: Video memory
- 16 10: DAC/LCDC
- 17 11: Display device
- 18 12: CRT display
- 19 13: Hard disk drive
- 20 14: ROM
- 21 15: Serial port
- 22 16: Parallel port
- 23 17: Timer
- 24 18: Communication adaptor
- 25 19: Floppy disk controller

- 1 20: Floppy disk drive
- 2 21: Audio controller
- 3 22: Amplifier
- 4 23: Loudspeaker
- 5 24: Microphone
- 6 25: IDE controller
- 7 26: CD-ROM
- 8 27: SCSI controller
- 9 28: MO
- 10 29: CD-ROM
- 11 30: Hard disk drive
- 12 31: DVD
- 13 32: DVD
- 14 100: System
- 15 DETAILED DESCRIPTION OF THE INVENTION:
- 16 Additional information embedding system
- 17 To achieve the above objects, according to the present
- 18 invention, a system for embedding additional information in
- 19 compressed audio data comprises:
- 20 (1) means for extracting MDCT (Modified Discrete Cosine
- 21 Transform) coefficients from the compressed audio data;
- 22 (2) means for employing the MDCT coefficients to calculate a
- 23 frequency component for the compressed audio data;

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- 1 (3) means for embedding additional information in the
- 2 frequency component obtained in a frequency domain;
- 3 (4) means for transforming into MDCT coefficients the
- 4 frequency component in which the additional information is
- 5 embedded; and
- 6 (5) means for using the MDCT coefficients, in which the
- 7 additional information is embedded, to generate compressed
- 8 audio data.
- 9 Additional information updating system
- 10 Further, according to the present invention, a system for
- 11 updating additional information embedded in compressed audio
- 12 data comprises:
- 13 (1) means for extracting MDCT coefficients from the
- 14 compressed audio data;
- 15 (2) means for employing the MDCT coefficients to calculate a
- 16 frequency component for the compressed audio data;
- 17 (3) means for detecting the additional information in the
- 18 frequency component that is obtained;
- 19 (3-1) means for changing, as needed, the additional
- 20 information for the frequency component;

- 1 (4) means for transforming into MDCT coefficients the
- 2 frequency component in which the additional information is
- 3 embedded; and
- 4 (5) means for using the MDCT coefficients, in which the
- 5 additional information is embedded, to generate compressed
- 6 audio data.
- 7 Additional information detection system
- 8 Further, according to the present invention, a system for
- 9 detecting additional information embedded in compressed
- 10 audio data comprises:
- 11 (1) means for extracting MDCT coefficients from the
- 12 compressed audio data;
- 13 (2) means for employing the MDCT coefficients to calculate a
- 14 frequency component for the compressed audio data; and
- 15 (3) means for detecting the additional information in the
- 16 frequency component that is obtained.
- 17 It is preferable that the means (2) calculate the frequency
- 18 component for the compressed audio data using a precomputed
- 19 table in which a correlation between MDCT coefficients and
- 20 frequency components is included.
- 21 It is also preferable that the means (4) transforms the

- 1 frequency component into the MDCT coefficients by using a
- 2 precomputed table that includes a correlation between MDCT
- 3 coefficients and frequency components.
- 4 In addition, it is preferable that the means (3) for
- 5 embedding the additional information in the frequency domain
- 6 divide an area for embedding one bit by the time domain, and
- 7 calculate a signal level for each of the individual obtained
- 8 area segments, while embedding the additional information in
- 9 the frequency domains in accordance with the lowest signal
- 10 level available for each frequency.
- 11 Correlation table generation method
- 12 According to the present invention, for at least one window
- 13 function and one window length employed for compressing
- 14 audio data, a method for generating a table including a
- 15 correlation between MDCT coefficients and frequency
- 16 components comprises:
- 17 (1) a step of generating a basis which is used for
- 18 performing a Fourier transform for a waveform along a time
- 19 axis;
- 20 (2) a step of multiplying a window function by a
- 21 corresponding waveform that is generated by using the basis;
- 22 (3) a step of performing an MDCT process, for the result
- 23 obtained by the multiplication of the window function, and

- 1 of calculating an MDCT coefficient; and
- 2 (4) a step of correlating the basis and the MDCT
- 3 coefficient. The example basis can be a sine wave and a
- 4 cosine wave.
- 5 Operation of additional information embedding system
- 6 The system for embedding additional information in
- 7 compressed audio data, first extracts compressed MDCT
- 8 coefficients from compressed digital audio data. Then, the
- 9 system employs MDCT coefficients sequence that have been
- 10 calculated and stored in a table in advance to obtain the
- 11 frequency component of the audio data. Thereafter, the
- 12 system employs the method for embedding additional
- 13 information in a frequency domain to calculate an embedded
- 14 frequency signal, and subsequently, the system employs the
- 15 table to transform the embedded frequency signal into a MDCT
- 16 coefficient, and adds the obtained MDCT coefficient to the
- 17 MDCT coefficient of the audio data. The resultant MDCT
- 18 coefficients are defined as new MDCT coefficients for the
- 19 audio data, and are again compressed; the resultant data
- 20 being regarded as watermarked digital audio data.
- 21 According to the method of the invention for embedding the
- 22 minimum data, a frame for the embedding therein of one bit
- 23 is divided at a time domain, a signal level is calculated
- 24 for each of the frame segments, and the upper embedding
- 25 limit is obtained in accordance with the lowest signal level

- 1 available for each frequency.
- 2 Operation performed for correlation table
- 3 A table for correlating the MDCT coefficient and the
- 4 frequency component is obtained in which representation of
- 5 each basis of a Fourier transformation relative to the MDCT
- 6 coefficient is calculated in advance in accordance with a
- 7 frame length (a window function and a window length). Thus,
- 8 an operation on the compressed audio data can be performed
- 9 directly.
- 10 The means for reducing the memory size that is required for
- 11 the correlation table employs the periodicity of the basis,
- 12 such as a sine wave or a cosine wave, to prevent the storage
- 13 of redundant information. Or, instead of storing in the
- 14 table the MDCT results obtained for the individual bases
- 15 using the Fourier transformation, each basis is divided into
- 16 several segments, and corresponding MDCT coefficients are
- 17 stored so that the memory size required for the table can be
- 18 reduced.
- 19 Operation of additional information detection system
- 20 The system of the invention employed detecting additional
- 21 information in compressed audio data, recovers coded MDCT
- 22 coefficients and employs the same table as is used for the
- 23 embedding system to perform a process equivalent to the
- 24 detection in the frequency domain and the detection of bit

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- 1 information and a code signal.
- 2 Operation of additional information updating system
- 3 The system of the invention, used for updating additional
- 4 information embedded in compressed audio data, recovers the
- 5 coded MDCT coefficients and employs the same method as the
- 6 detection system to detect a signal embedded in the MDCT
- 7 coefficients. Only when the strength of the embedded signal
- 8 is insufficient, or when a signal that differs from a signal
- 9 to be embedded is detected and updating is required, the
- 10 same method is employed as that used by the embedding system
- 11 to embed additional information in the MDCT coefficients.
- 12 The newly obtained MDCT coefficients are thereafter recorded
- 13 so that they can be employed as updated digital audio data.
- 14 Preferred Embodiment
- 15 First, definitions of terms will be given before the
- 16 preferred embodiment of the invention is explained.
- 17 Sound compression technique
- 18 Compressed data for the present invention are electronic
- 19 compressed data for common sounds, such as voices, music and
- 20 sound effects. The sound compression technique is well
- 21 known as MPEG1 or MPEG2. In the specification, this
- 22 compression technique is generally called the sound
- 23 compression technique, and the common sounds are described

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- 1 as sound or audio.
- 2 * Compressed state
- 3 The compressed state is the state wherein the amount of
- 4 audio data is reduced by the target sound compression
- 5 technique, while deterioration of the sound is minimized.
- 6 * Non-compressed state
- 7 The non-compressed state is a state wherein an audio
- 8 waveform, such as a WAVE file or an AIFF file, is described
- 9 without being processed.
- 10 * Decode the compressed state
- 11 This means "convert from the compressed state of the audio
- 12 data to the non-compressed state." This definition is also
- 13 applied to "shifting to the non-compressed state."
- 14 * MDCT transform (Modified Discrete Cosine Transform)
- 15 Equation 1
- 16 [All the equations are tabulated at the end of the text of
- 17 this description, just before the claims.]
- 19 index along the time axis.

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- 1 Mk denotes a MDCT coefficient, and k is an integer of from 0
- 2 to (N/2)-1, and denotes an index indicating a frequency.
- 3 In the MDCT transform, the sequence X0 to X(N-1) along the
- 4 time axis are transformed into the sequence M0 to M((N/2)-1)
- 5 along the frequency axis. While the MDCT coefficient
- 6 represents one type of frequency component, in this
- 7 specification, the "frequency component" means a coefficient
- 8 that is obtained as a result of the DFT transform.
- 9 * DFT transform (Discrete Fourier Transform)
- 10 Equation 2
- 12 an index along the time axis.
- 13 Rk denotes a real number component (cosine wave component);
- 14 Ik denotes an imaginary number component (sine wave
- 15 component); and k is an integer of from 0 to (N/2)-1, and
- 16 denotes an index indicating a frequency. The discrete
- 17 fourier transform is a transformation of the sequence X0 to
- 18 $\times (N-1)$ along the time axis into the sequences R0 to
- 19 R((N/2)-1), and I0 to I((N/2)-1) along the frequency axis.
- 20 In this specification, "frequency component" is the general
- 21 term for the sequences Rk and Ik.
- 22 * Window function

- 1 This function is to be multiplied by the sample value before
- 2 the MDCT is performed. Generally, the sine function or the
- 3 Kaiser function is employed.
- 4 * Window length
- 5 The window length is a value that represents the shape or
- 6 length of a window function to be multiplied with data in
- 7 accordance with the characteristic of the audio data, and
- 8 that indicates whether the MDCT should be performed for
- 9 several samples.
- 10 Fig. 1 is a block diagram showing the processing performed
- 11 by an apparatus for directly embedding additional
- 12 information in compressed audio data. A block 110 is a
- 13 block for extracting MDCT coefficients sequence from
- 14 compressed audio data that are entered. A block 120 is a
- 15 block for employing the extracted MDCT coefficients to
- 16 calculate the frequency component of the audio data. A
- 17 block 130 is a block for embedding additional information in
- 18 the obtained frequency component of a frequency domain. A
- 19 block 140 is a block for transforming the frequency
- 20 component using the additional information embedded in an
- 21 MDCT coefficient. And finally, a block 150 is a block for
- 22 generating compressed audio data by using the MDCT
- 23 coefficient obtained by the block 140.
- 24 The blocks 120 and 130 employ a correlation table for the

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- 1 MDCT coefficient and the frequency to perform a fast
- 2 transform. In this invention, the representations of the
- 3 bases of the Fourier transform in the MDCT domain are
- 4 entered in advance in the table, and are employed for the
- 5 individual embedding, detection and updating systems. An
- 6 explanation will now be given for the correlation table for
- 7 the MDCT coefficient and the frequency and the generation
- 8 method therefor, the systems used for embedding, detecting
- 9 and updating compressed audio data, and other associated
- 10 methods.
- 11 Correlation table for MDCT coefficients and frequency
- 12 components
- 13 Audio data must be transformed into a frequency domain in
- 14 order to employ an auditory psychological model for
- 15 embedding calculation. However, a very extended calculation
- 16 time is required to perform inverse transformations, for the
- 17 audio data that are represented as MDCT coefficients, and to
- 18 perform the Fourier transforms for audio data at the time
- 19 domain. Thus, a correlation between the MDCT coefficients
- 20 and the frequency components is required.
- 21 If the audio data are compressed by performing the MDCT for
- 22 a constant number of samples without a window function, the
- 23 MDCT employs the cosine wave with a shifted phase as a
- 24 basis. Therefore, the difference from a Fourier transform
- 25 consists only of the shifting of a phase, and a preferable
- 26 correlation can be expected between the MDCT domain and the

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- 1 frequency domain. However, to obtain improved tone quality,
- 2 the latest compression technique changes the shape or the
- 3 length of the window function to be multiplied (hereinafter
- 4 refereed to as a window length) in accordance with the
- 5 characteristic of the audio data. Thus, a simple
- 6 correlation between a specific frequency for the MDCT and a
- 7 specific frequency for a Fourier transform can not be
- 8 obtained, and since the correlation can not be acquired
- 9 through calculation, it must be stored in a table.
- 10 Fig. 2 is a diagram showing window length and window
- 11 function examples. While this invention can be applied for
- 12 various compressed data standards, in this embodiment, the
- 13 MPEG2 standards are employed. For MPEG2 AAC (Advanced Audio
- 14 Coding), for example, a window function normally having a
- 15 window length of 2048 samples is multiplied to perform the
- 16 MDCT. For a portion where sound is drastically altered, a
- 17 window function having a window length of 256 samples is
- 18 multiplied to perform the MDCT, so that a type of
- 19 deterioration called pre-echo is prevented. A normal frame
- 20 for which 2048 samples is a unit is called an
- 21 ONLY LONG_SEQUENCE, and is written using 1024 MDCT
- 22 coefficients that are obtained from one MDCT process. A
- 23 frame for which 256 samples is a unit is called an
- 24 EIGHT_SHORT_SEQUENCE, and is written using eight pairs of
- 25 MDCT 128 coefficients that are obtained by repeating the
- 26 MDCT eight times, for 256 samples each time, with each frame
- 27 half overlapping its adjacent frame. Further, asymmetric
- 28 window functions called a LONG_START_SEQUENCE and a

- 1 LONG STOP SEQUENCE are also employed to connect the above
- 2 frames.
- 3 Fig. 3 is a diagram showing the correlation between the
- 4 window functions and the MDCT coefficients sequence. For
- 5 the MPEG2 AAC, the window functions are multiplied by the
- 6 audio data along the time axis, for example, in the order
- 7 indicated by the curves in Fig. 3, and the MDCT coefficients
- 8 are written in the order indicated by the thick arrows.
- 9 When the window length is varied, as in this example, the
- 10 bases of a Fourier transform can not simply be transformed
- 11 into a number of MDCT coefficients.
- 12 Therefore, to embed additional information, the correlation
- 13 table of this invention does not depend on the window
- 14 function (a signal added during the additional information
- 15 embedding process should not depend on a window function
- 16 when the signal is decompressed and developed along the time
- 17 axis). Therefore, when an embedding method is employed that
- 18 depends on the shape of the window function and the window
- 19 length, the embedding and the detection of the compressed
- 20 audio data can be performed, and the window function that is
- 21 used can be identified when the data are decompressed.
- 22 The correlation table of the invention is generated so that
- 23 frames in which additional information is to be embedded do
- 24 not interfere with each other. That is, in order to embed
- 25 additional information, the MDCT window must be employed as
- 26 a unit, and when the data are developed along the time axis,

- 1 one bit must be embedded in a specific number of samples,
- 2 which together constitute one frame. Since for the MDCT,
- 3 target frames for the multiplication of a window overlap
- 4 each other 50%, a window that extends over a plurality of
- 5 frames is always present (a block 3 in Fig. 4 corresponds to
- 6 such a window). When additional information is simply
- 7 embedded in one of these frames, it affects the other
- 8 frames. And when data embedding is not performed, the data
- 9 embedding intensity is reduced, as is detection efficiency.
- 10 Signals indicating different types of additional information
- 11 are embedded in the first and the second halves of a frame.
- 12 The correlation table is employed when a frequency component
- 13 is to be calculated using the MDCT coefficient to embed
- 14 additional information, when an embedded signal obtained at
- 15 the frequency domain is to be again transformed into an MDCT
- 16 coefficient, and when a calculation corresponding to a
- 17 detection in a frequency domain is to be performed in the
- 18 MDCT domain. Since the detection and the embedding of a
- 19 signal are performed in order during the updating process,
- 20 all the transforms described above are employed in the
- 21 updating process.
- 22 Method for generating a correlation table when the length of
- 23 a window function is unchanged
- 24 First, an explanation will be given for the table generation
- 25 method when a window length is constant, and for the
- 26 detection and embedding methods that use the table. These

- 1 methods will be extended later for use by a plurality of
- 2 window lengths. Assume that the window function is
- 3 multiplied along the time axis by audio data consisting of N
- 4 samples and the MDCT is performed to obtain N/2 MDCT
- 5 coefficients, and that N/2 MDCT coefficients are employed
- 6 and written as one block (i.e., a constant window length is
- 7 defined as N samples). Hereinafter, if not specifically
- 8 noted, the term "block" represents N/2 MDCT coefficients.
- 9 The audio data along the time axis that correspond to two
- 10 sequential blocks are those where there is a 50%, i.e., N/2
- 11 samples, overlap.
- 12 The target of the present invention is limited to an
- 13 embedding ratio for the embedding of one bit in relative
- 14 samples integer times N/2. In this embodiment, the number
- 15 of samples required along the time axis to embed one bit is
- 16 defined as $n\times N/2$, which is called one frame. Due to the
- 17 previously mentioned 50% overlapped property there is also a
- 18 block that is extended across two sequential frames along
- 19 the time axis. Fig. 4 is a specific diagram showing two
- 20 frames extended along the time axis when n=2 that correspond
- 21 to five blocks in the MDCT domain. The audio data along the
- 22 time axis are shown in the lower portion in Fig. 4, the MDCT
- 23 coefficients sequence are shown in the upper portion, and
- 24 elliptical arcs represent the MDCT targets. Block 3 is a
- 25 block extending half way across Frame 1 and Frame 2.
- 26 Since the embedding operation is performed for the
- 27 independent frames, the correlation between the frequency

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- 1 component and the MDCT coefficient for each frame need only
- 2 be required for the table. In other words, adjacent frames
- 3 in which embedding is performed should not affect each
- 4 other. Therefore, for each basis of a Fourier transform
- 5 having a cycle of $N/(2\times m)$, the MDCT coefficients sequence
- 6 obtained using the following methods are employed to prepare
- 7 a table. In this case, m is an integer equal to or smaller
- 8 than N/2. Fig. 5 is a diagram showing a sine wave for n=2
- 9 and m=1.
- 10 There are n+1 blocks present that are associated with one
- 11 frame, and the first and the last blocks also extend into
- 12 the respective succeeding and preceding frames (blocks 1 and
- 13 3 in Fig. 5). Thus, assume a waveform (the thick line
- 14 portion in Fig. 5) is obtained by connecting N/2 samples
- 15 having a value of 0 before and after the basis waveform that
- 16 has an amplitude of 1.0 and a length equivalent to one
- 17 frame. When a window function (corresponding to an
- 18 elliptical arc in Fig. 5) is multiplied by N samples, while
- 19 50% of the first part of the waveform is overlapped, and the
- 20 MDCT is performed, this waveform can be represented by using
- 21 the MDCT coefficients. If the IMDCT is performed for the
- 22 obtained MDCT coefficients sequence, the preceding and
- 23 succeeding N/2 samples have a value of 0.
- 24 Fig. 6 is a diagram showing an example wherein additional
- 25 information is embedded in adjacent frames. When samples
- 26 having a value of 0 are added as shown in Fig. 6, the
- 27 interference produced by embedding performed in adjacent

- 1 frames can be prevented. In the data detection process and
- 2 the frequency component calculation process, detection
- 3 results and frequency components can be obtained that are
- 4 designated for a pertinent frame and that are not affected
- 5 by preceding and succeeding frames. If a value of 0 is not
- 6 compensated for, adjacent frames affect each other in the
- 7 embedding and detection process.
- 8 The processing performed to prepare the table is as follows.
- 9 Step 1: First, calculations are performed for a cosine wave
- 10 having a cycle of $N/2\times n/k$, an amplitude of 1.0 and a length
- 11 of $N/2\times n$. This cosine wave corresponds to the k-th basis
- 12 when a Fourier transform is to be performed for the N/2xn
- 13 samples.

14
$$f(x) = cos(2\pi/(N/2\times n/k)\times x)$$
 $(0 \le x < N/2\times n)$

$$15 = \cos(4k\pi/(N\times n)\times x)$$

- 16 Step 2: N/2 samples having a value of 0 are compensated for
- 17 at the first and the last of the waveform (Fig. 5).

18
$$g(y) = 0$$
 $(0 \le y < N/2)$

19
$$f(y-N/2)$$
 $(N/2 \le y < N/2 \times (n+1))$

20 0
$$(N/2 \times (n+1) \le y < N/2 \times (n+2))$$

- 1 Step 3: The samples $N/2\times(b-1)$ th to $N/2\times(b+1)$ th are
- 2 extracted. Here b is an integer of from 1 to n+1, and for
- 3 all of these integers the following process is performed.
- 4 $h_b(z) = g(z+N/2\times(b-1))$ (0\leq z<N)
- 5 Step 4: The results are multiplied by a window function.
- 6 $h_b(z) = h_b(z) \times win(z)$ (0\leq z<N, win(z) is a window
- 7 function)
- 8 Step 5: The MDCT process is performed, and the obtained N/2
- 9 MDCT coefficients are defined as vectors $V_{\text{r,b,k}}$.
- 10 $V_{r,b,k} = MDCT(h_b(z))$
- 11 Since the MDCT transform is an orthogonal transform and each
- 12 basis of a Fourier transform is a linear independence, $V_{\text{r,b,k}}$
- 13 are orthogonal for a k having a value of 1 to N/2.
- 14 Step 6: $V_{r,b,k}$ is obtained for all the combinations (k, b),
- 15 and each matrix $T_{r,b}$ is formed.
- 16 $T_{r,b} = (V_{r,b,1}, V_{r,b,2}, V_{r,b,3}, ... V_{r,b,N/2})$
- 17 The vector that is obtained for a sine wave using the same
- 18 method is defined as vi, b, k, and the matrix is defined as
- 19 Ti, b. Each sequence is an MDCT coefficient sequence that
- 20 represents the sine wave of a value of 1. Since there are 1

- 1 to n+1 blocks, $2 \times (n+1)$ matrixes are obtained.
- 2 Transform from a frequency domain into an MDCT domain
- 3 Assume that the audio data in the frequency domain are
- 4 represented as R + jI, where j denotes an imaginary number
- 5 element, R denotes a real number element and I is the N/2th
- 6 order real number vector that represents an imaginary number
- 7 element. The k element corresponds to a basis having a
- 8 cycle of $(N/2) \times n/k$ samples. The MDCT coefficient sequence
- 9 Mb is obtained as the sum of the vectors of MDCT
- 10 coefficients sequence, which is obtained by transforming
- 11 each frequency component separately into an MDCT domain, and
- 12 can be represented as $M_b = T_{r,b} + T_{i,b}I$. In this case, b is an
- 13 integer of from 1 to n+1, and corresponds to each block. M1
- 14 and Mn+1 are MDCT coefficients sequence for a block that
- 15 extends across portions of adjacent frame.
- 16 Transform from an MDCT domain into a frequency domain
- 17 Here, vi,b,k and the vr,b,k are orthogonal to each other and
- 18 form an MDCT domain. Thus, when a specific MDCT coefficient
- 19 sequence is given, and when the inner product is calculated
- 20 for the MDCT coefficient sequence and vr,b,k or vi,b,k, the
- 21 element in the corresponding direction of the Mb can be
- 22 obtained that represents respectively a real number element
- 23 and/or an imaginary number element in the frequency domain.
- 24 The MDCT coefficients sequence for (n+1) blocks associated
- 25 with one frame are collectively processed to obtain the

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- 1 frequency component for the pertinent frame.
- 2 Equation 3
- 3 Correlation table generation method when a window function
- 4 is changed in audio data
- 5 Assume that the types of window functions that could be
- 6 employed for compression are listed. All the window lengths
- 7 are dividers having a maximum window length of N. For a
- 8 block having an N/W (W is an integer) sample window length,
- 9 assume that the MDCT is repeated for the N/W sample W times,
- 10 with 50% overlapping, and that as a result W pairs of N/(2W)
- 11 MDCT coefficients, i.e., a total of N/2 coefficients, are
- 12 written in the block. Further, assume that in the first
- 13 MDCT process N/W samples beginning with the "offset" sample
- 14 in the block are transformed. For example, where for the
- 15 EIGHT_SHORT_SEQUENCE of the MPEG2 AAC, N=2048, W=8 and
- 16 offset=448. As a result of repeating the eight MDCT
- 17 processes for 256 samples with 50% overlapping, eight pairs
- 18 of 128 MDCT coefficients are written along the time axis
- 19 (see Figs. 2 and 3).
- 20 Table generation method
- 21 The table for the window length N/W is generated as follows.
- 23 Step 1: The same as when the length of the window function

- 1 is unchanged.
- 2 Step 2: The same as when the length of the window function
- 3 is unchanged.
- 4 Step 3: The N/W sample corresponding to the W-th window is
- 5 extracted. W is an integer of from 1 to W. b is an integer
- 6 of from 1 to n+1. The following processing must be
- 7 performed for all the combinations of b and w.
- 8 $h_{b,w}(z) = g(z+N/2\times(b-1)+N/2/W\times w+offset)$
- $9 \qquad \qquad (0 \le z < N/W)$
- 10 Step 4: The results are multiplied by a window function.
- 11 $h_{b,w}(z) = h_{b,w}(z) \times win(z)$ (0\leq z<N/W: win(z) is a
- 12 window function)
- 13 Step 5: The MDCT process is performed, and the obtained
- 14 N/(2W) MDCT coefficients are defined as vectors $V_{r,b,k,w}$.
- 15 $V_{r,b,k,w} = MDCT(h_{b,w}(z))$
- 16 Step 6: $v_{r,b,k,w}$ are arranged to define $v_{r,b,k}$.
- 17 When $v_{r,b,k,w}$ is obtained for all the "w"s having a value of 1
- 18 to W, they are arranged vertically to obtain vector $V_{r,b,k}$.

- 1 Fig. 7 is a diagram showing the portion of a basis for
- 2 which, with n=2, b=2, k=1 and W=8, the MDCT process has been
- 3 performed to obtain the coefficients $v_{r,2,1,w}$.
- 4 Step 7: The coefficients $V_{r,b,k}$ are obtained for all the
- 5 combinations (k, b), and the coefficients $v_{r,b,k}$ for k having
- 6 values of 1 to N/2 are arranged horizontally to constitute
- 7 $T_{\text{W.r.b.}}$
- 8 Since each $v_{r,b,k,w}$ is a vector of N/(2w) rows by one column,
- 9 this matrix is a square matrix of N/2 rows by N/2 columns.
- 10 Each column illustrates how a cosine wave having a value of
- 11 1 is represented as the MDCT coefficients sequence in the
- 12 b-th block having a window length of N/W. Similarly, the
- 13 matrix TW, i, b is obtained in the sine wave. Since from 1 to
- 14 n+1 block numbers b are provided, for this window length, 2
- 15 \times (n+1) matrixes are obtained. In addition, the table is
- 16 prepared in accordance with the window length and the types
- 17 of window functions.
- 18 Transform from the frequency domain to the MDCT domain
- 19 The difference from a case where only one type of window
- 20 length is employed is that block information is read from
- 21 compressed audio data and that a different matrix is
- 22 employed in accordance with the window function that is used
- 23 for each block. Since the matrix is varied for each block,
- 24 the MDCT coefficient sequence Mb is adjusted in order to
- 25 cope with the window function and the window length that are

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- 1 employed. The waveform, which is obtained when the IMDCT is
- 2 performed for the MDCT coefficient sequence Mb in the time
- 3 domain, and the frequency component, which is obtained by
- 4 performing a Fourier transform in the frequency domain, do
- 5 not depend on the window function and the window length.
- 6 The MDCT coefficient sequence Mb is obtained using Mb =
- 7 $T_{w,r,b}R + T_{w,1,b}I$.
- 8 Transform from the MDCT domain to the frequency domain
- 9 When $T_{w,r,b}$ is employed instead of $T_{r,b}$, the transform in the
- 10 frequency domain can be performed in the same manner. When
- 11 the matrix is changed in accordance with the window function
- 12 and the window length, a true frequency component can be
- 13 obtained that does not depend on the window function and the
- 14 window length.
- 15 Equation 4
- 16 Method for reducing a memory capacity required for the table
- 17 Since the matrix has a size of $(N/2)\times(N/2)$, the table
- 18 generated by this method is constituted by 2 \times (n+1) \times (N/2)
- $19 \times (N/2) = (n+1) \times N/2/2$ MDCT coefficients (floating-point
- 20 numbers). However, since the contents of this table tend to
- 21 be redundant, the memory capacity that is actually required
- 22 can be considerably reduced.
- 23 Method 1: method for using the periodicity of the basis.

- 1 The periodicity of the basis can be employed as one method.
- 2 According to this method, since several $V_{r,b,k}$ are identical,
- 3 this portion is removed.
- 4 When m is an integer, the cosine wave that is $N/2 \times m$ samples
- 5 ahead is represented as
- $6 f(x+N/2\times m) = cos(4k\pi/(N\times n)\times(x+N/2\times m))$
- $7 = \cos(4k\pi/(N\times n)\times x + 4k\pi/(N\times n)\times N/2\times m)$
- $= \cos (4k\pi/(Nxn)xx + 2\pi kxm/n).$
- 9 Therefore, in case a where (kxm)/n is an integer,
- 10 $f(x+N/2\times m) = f(x)$ (limited to a range $0 \le x \le N/2 \times (n-m)$)
- 11 $g(y+N/2\times m) = g(y)$ (limited to a range $N/2 \le y \le N/2 \times (n-m+1)$).
- 12 Thus,
- 13 $h_{b+m}(z) = h_b(z)$ (limited to a range $2 \le b \le n-m$),
- 14 and
- 15 $V_{r,b+m,k} = V_{r,b,k}$ (limited to a range $2 \le b \le n-m$)
- 16 is obtained. The range is limited because of the range

- In case b where (kxm)/n is an irreducible fraction that can 2
- be represented by integer/2, 3
- $f(x+N/2\times m) = -f(x)$
- 5 And
- $h_{b+m}(z) = -h_b(z)$. 6
- 7 Thus,

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- $V_{r,b+m,k} = -V_{r,b,k}.$ 8
- The range limitation is the same as it is for case a. 9
- In case c where (kxm)/n is an irreducible fraction that can 10
- be represented by $(4 \times integer + 1)/4$, 11
- $f(x+N/2\times m) = cos(4k\pi/(N\times n)\times x + \pi(even number+1/2))$ 12
- = $-\sin(4k\pi/(N\times n)\times x)$. 13
- 14 Thus,
- $V_{r,b+m,k} = -V_{l,b,k}$ 15
- In case d where (kxm)/n is an irreducible fraction that can 16

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- 1 be represented by $(4\times integer+3)/4$,
- 2 $f(x+N/2\times m) = cos(4k\pi/(N\times n)\times x + \pi(odd number+1/2))$
- $= \sin(4k\pi/(N\times n)\times x).$
- 4 Thus,
- $5 \quad V_{r,b+m,k} = V_{i,b,k}.$
- 6 The range limitation is the same as it is for case a.
- 7 Therefore, $V_{r,b+m,k}$, which establishes conditions a to d, can
- 8 be replaced by another vector, and this is applied to $V_{i,b\;k}$.
- 9 Thus, instead of storing the matrixes $T_{\text{r,b}}$ and $T_{\text{1,b}}$ being
- 10 unchanged, only the following minimum elements need be
- 11 stored. The following minimum elements are as follows.
- 12 * vectors $V_{r,b,\,\kappa}$ and $V_{i,b,k}$ that do not establish the conditions
- 13 a to d
- 14 * information concerning the positive or negative sign that
- 15 is to be added to a vector that is to be used for each
- 16 column in the matrixes $T_{\text{r,b}}$ and $T_{\text{i,b}}.$
- 17 For the actual transform between the MDCT domain and the
- 18 frequency domain, the vectors $V_{\text{r,b,k}}$ and $V_{\text{i,b,k}}$ are employed
- 19 instead of the columns in the matrixes $T_{\text{r,b}}$ and $T_{\text{i,b}}$ to perform
- 20 a calculation equivalent to the matrix operation. The

- 1 transform from the frequency domain to the MDCT domain is
- 2 represented as follows.
- 3 Equation 5
- 4 Another appropriate vector is employed for a portion wherein
- 5 a vector is standardized. The transform from the MDCT
- 6 domain to the frequency domain is performed by obtaining the
- 7 following inner product for each frequency component. The
- 8 following equation is obtained by separating the equation
- 9 $\,$ used for the matrixes $T_{\text{r},\text{b}}$ and $T_{\text{1},\text{b}}$ into its individual
- 10 components.
- 11 Equation 6
- 12 Due to the vector standardization, the required memory
- 13 capacity depends on "n" to a degree. For example, since
- 14 only the condition a is established when n=3, the required
- 15 memory capacity is reduced only 8.3%, while when n=4, it is
- 16 reduced 40%.
- 17 Since the same relation exists between hb and w as when only
- 18 one type of window function is provided in a case where the
- 19 window function is varied, the above standardization can be
- 20 employed unchanged, and when the same condition is
- 21 established, the following equation is obtained.
- 22 Equation 7

- 1 Method 2: method for separating the basis into preceding and
- 2 succeeding segments.
- 3 Furthermore, the linearity of the MDCT is employed to
- 4 separate the basis of a Fourier transform into individual
- 5 segments, and the MDCT coefficients sequence obtained by the
- 6 transform are used to form a table. Then, the application
- 7 range of the above method 1 can be expanded. Actually, the
- 8 sum of the vectors of the MDCT coefficients sequence that
- 9 are stored in the table is employed to represent the basis.
- 10 Fig. 8 is a diagram showing an example wherein a basis is
- 11 separated.
- 12 First, a waveform (thick line on the left in Fig. 8) is
- 13 divided into the first N/2 sample and the last N/2 sample
- 14 for each block. To perform an MDCT for the first N/2
- 15 sample, a waveform having a value of 0 is compensated for by
- 16 the N/2 sample (in the middle in Fig. 8). To perform an
- 17 MDCT for the last N/2 sample, a wave form having a value of
- 18 0 is compensated for by the N/2 sample (on the right in Fig.
- 19 8). In this example, the MDCT is performed for the first
- 20 (last) half of the waveform, and the obtained MDCT
- 21 coefficients sequence are represented by $V_{\text{fore},r,b,k}$ $(V_{\text{back},r,b,k})$.
- 22 Since the MDCT possesses linearity, the original MDCT
- 23 coefficient sequence $V_{r,b,k}$ is equal to the sum of the vectors
- $V_{\text{fore,r,b,k}}$ and $V_{\text{back,r,b,k}}$.
- 25 When the basis is separated in this manner, $V_{\text{fore},r,b,k}$ and
- $V_{\text{back},r,b,k}$ can be used in common even for the portion wherein

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- 1 $V_{r,b,k}$ can not be standardized using method 1. For example,
- 2 in Fig. 5, method 1 can not be applied for Block 1 because
- 3 b=1. However, if each block is separated into first and
- 4 last segments, the signs are merely inverted for the MDCT
- 5 coefficient sequence $V_{\text{back},r,1,k}$ for Block 1 and the MDCT
- 6 coefficient sequence $V_{\text{back},r,2,k}$ for Block 2. Therefore, one of
- 7 the MDCT coefficients sequence need not be stored. This can
- 8 also be applied for $V_{\text{fore},r,2,k}$, for Block 2, and $V_{\text{fore},r,3,k}$, for
- 9 Block 3. $V_{\text{fore},r,1,k}$, for Block 1, and $V_{\text{fore},r,3,k}$, for Block 3 are
- 10 always zero vectors.
- 11 The processing for generating a table using the above method
- 12 is as follows.
- 13 Step 1: The same as when the basis is not separated into
- 14 first and second segments.
- 15 Step 2: The same as when the basis is not separated into
- 16 first and second segments.
- 17 Step 3: First, the "fore" coefficients are prepared. The
- 18 $(N/2\times(b-1))$ -th to the $(N/2\times b)$ -th coefficients are extracted,
- 19 and the N/2 sample having a value of 0 is added after them.
- 20 $h_{fore,b}(z) = g(z+N/2\times(b-1)) (0 \le z < N/2)$
- 21 0 $(N/2 \le z < N)$
- 22 Step 4: A window function is multiplied.

- $1 \quad h_{fore,b}(z) = h_{fore,b}(z) \times win(z)$
- 2 $(0 \le z < N, win(z) \text{ is a window function})$
- 3 Step 5: The MDCT process is performed, and the obtained N/2
- 4 MDCT coefficients are defined as vector $V_{fore,r,b,k}$.
- 5 $V_{fore,r,b,k} = MDCT(h_{fore,b}(z))$.
- 6 Step 6: Next, the "back" coefficients are prepared. The
- 7 (N/2xb)-th to the (N/2x(b+1))-th coefficients are extracted,
- 8 and the N/2 sample having a value of 0 is added before them.
- $9 h_{back,b}(z) = 0 (0 \le z < N/2)$
- 10 $g(z+N/2\times(b-1))$ $(N/2\leq z< N)$
- 11 Step 7: A window function is multiplied.
- 12 $h_{back,b}(z) = h_{back,b}(z) \times win(z)$
- 13 $(0 \le z < N, win(z) \text{ is a window function})$
- 14 Step 8: The MDCT process is performed, and the obtained N/2
- 15 MDCT coefficients are defined as vector $V_{\text{back},r,b,k}$.
- 16 $V_{back,r,b,k} = MDCT(h_{back,b}(z))$.

- 1 Step 9: $V_{\text{fore},r,b,k}$ and $V_{\text{back},r,b,k}$ are calculated for all the
- 2 combinations (k,b), and the matrixes $T_{\text{fore},r,b}$ and $T_{\text{back},r,b}$ are
- 3 formed.
- 4 $T_{\text{fore,r,b}} = (V_{\text{fore,r,b,1}}, V_{\text{fore,r,b,2}}, \dots V_{\text{fore,r,b,N/2}})$
- 5 $T_{back,r,b} = (V_{back,r,b,1}, V_{back,r,b,2}, ..., V_{back,r,b,N/2})$
- 6 In accordance with the linearity of the MDCT,
- $7 \quad V_{r,b,k} = V_{fore,r,b,k} + V_{back,r,b,k},$
- 8 and
- $9 \quad T_{r,b} = T_{fore,r,b} + T_{back,r,b}.$
- 10 In accordance with this characteristic, for the transform
- 11 between the MDCT domain and the frequency domain, only an
- 12 operation equivalent to the operation performed using the
- 13 $T_{r,b}$ need be performed by using $T_{fore,r,b}$ and $T_{back,r,b}$.
- 14 The periodicity of the basis is employed under these
- 15 definitions,
- 16 in case a where $(k \times m)/n$ is an integer, and under the
- 17 condition where b+m=n+1,
- $h_{fore,n+1}(z) = h_{fore,b}(z)$ is established. This is because the
- 19 second half of $h_{\text{fore,b}}(z)$ has a value of 0. Thus, the

```
application range for the following equation is expanded,
      2
           and
            h_{fore,b+m}(z) == h_{fore,b}(z)
      3
                                     (limited to a range of 2 \le b \le n-m+1).
      4
      5
            Thus,
            V_{\text{fore,r,b+m,k}} == V_{\text{fore,r,b,k}}
Santa R. of H. adjung at Bron. R. Ronelle R. Bron. R. R. and R. B. dinon. Stanford Stanford Stanford Stanford
                                     (limited to a range of 2 \le b \le n-m+1),
      7
            and the portions used in common are increased. For V_{\text{back},r,b,k},
       8
       9
The R. R. W. Comb. Comb. Comb. R. Co. Start Start Start Start Start Start Start Start
            h_{back,m+1(z)} == h_{back,1(z)}
     10
            is established even under the condition where b=1.
      11
            because the first half of l(z) has a value of zero.
      12
            application range for the following equation is expanded,
      13
      14
             and
```

(limited to a range of $1 \le b \le n-m$).

17 Therefore,

15

16

 $h_{back,b+m}(z) == h_{back,b}(z)$

- 1 $V_{back,r,b+m,k} == V_{back,r,p,k}$
- 2 (limited to a range of $1 \le b \le n-m+1$),
- 3 and the portions used in common are increased. The same
- 4 range limitation is provided for the cases b, c and d.
- 5 Method 3: approximating method
- 6 The final method for reducing the table involves the use of
- 7 an approximation. Among the MDCT coefficients sequence that
- 8 correspond to one basis waveform of a Fourier transform, an
- 9 MDCT coefficient that is smaller than a specific value can
- 10 approximate zero, and no actual problem occurs. A threshold
- 11 value used for the approximation is appropriately selected
- 12 by a trade off between the transform precision and the
- 13 memory capacity. When the individual systems are so
- 14 designed that they do not perform a matrix calculation for
- 15 the portion that approximates zero, the calculation time can
- 16 also be reduced.
- 17 Furthermore, when all the coefficients, including large
- 18 coefficients, approximate rational numbers, which are then
- 19 quantized, the coefficients can be stored as integers, not
- 20 as floating-point numbers, so that a savings in memory
- 21 capacity can be realized.
- 22 Correlation table generator

- 1 Information concerning the window is received, and the table
- 2 is generated and output. As well as the method for
- 3 generating the correlation table, the information concerning
- 4 the window includes the frame length N, the length n of a
- 5 block corresponding to the frame, the offset of the first
- 6 window, the window function, and "W" for regulating the
- 7 window length. Basically, the number of tables that are
- 8 generated is equivalent to the number of window types used
- 9 in the target sound compression technique.
- 10 Additional information embedding system
- 11 Fig. 9 is a block diagram illustrating an additional
- 12 information embedding system according to the present
- 13 invention. An MDCT coefficient recovery unit 210 recovers
- 14 sound MDCT coefficients sequence, and window and other
- 15 information from compressed audio data that are entered.
- 16 These data are extracted (recovered) using Huffmann
- 17 decoding, inverse quantization and a prediction method,
- 18 which are designated in the compressed audio data. An
- 19 MDCT/DFT transformer 230 receives the sound MDCT
- 20 coefficients sequence and the window information that are
- 21 obtained by the MDCT coefficient recovery unit 210, and
- 22 employs a table 900 to transform these data into a frequency
- 23 component. A frequency domain embedding unit 250 embeds
- 24 additional information in the frequency component that is
- 25 obtained by the MDCT/DFT transformer 230.
- 26 In accordance with the window information extracted by the

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- 1 MDCT coefficient recovery unit 210, a DFT/MDCT transformer
- 2 240 employs the table 900 to transform, into MDCT
- 3 coefficients sequence, the resultant frequency components
- 4 that are obtained by the frequency domain embedding unit
- 5 250. Finally, an MDCT coefficient compressor 220 compresses
- 6 the MDCT coefficients obtained by the DFT/MDCT transformer
- 7 240, as well as the window information and the other
- 8 information that are extracted by the MDCT coefficient
- 9 recovery unit 210. The compressed audio data are thus
- 10 obtained. The prediction method, the inverse quantization
- 11 and the Huffmann decoding, which are designated in the
- 12 window information and the other information, are employed
- 13 for the data compression. Through this processing, the
- 14 additional information is embedded so it corresponds to the
- 15 operation of the frequency component, and so that even after
- 16 decompression additional information can be detected using
- 17 the conventional frequency domain detection method.
- 18 Additional information detection system
- 19 Fig. 10 is a block diagram illustrating an additional
- 20 information detection system according to the present
- 21 invention. An MDCT coefficient recovery unit 210 recovers
- 22 sound MDCT coefficients sequence, window information and
- 23 other information from compressed audio data that are
- 24 entered. These data are extracted (recovered) using
- 25 Huffmann decoding, inverse quantization and a prediction
- 26 method, which are designated in the compressed audio data.
- 27 An MDCT/DFT transformer 230 receives the sound MDCT

- 1 coefficients sequence and the window information that are
- 2 obtained by the MDCT coefficient recovery unit 210, and
- 3 employs a table 900 to transform these data into frequency
- 4 components. Finally, a frequency domain detector 310
- 5 detects additional information in the frequency components
- 6 that are obtained by the MDCT/DFT transformer 230, and
- 7 outputs the additional information.
- 8 Additional information updating system
- 9 Fig. 11 is a block diagram illustrating an additional
- 10 information updating system according to the present
- 11 invention.
- 12 An MDCT coefficient recovery unit 210 recovers sound MDCT
- 13 coefficients sequence, window information and other
- 14 information from compressed audio data that are entered.
- 15 These data are extracted (recovered) using Huffmann
- 16 decoding, inverse quantization and a prediction method,
- 17 which are designated in the compressed audio data.
- 18 An MDCT/DFT transformer 230 receives the sound MDCT
- 19 coefficients sequence and the window information that are
- 20 obtained by the MDCT coefficient recovery unit 210, and
- 21 employs a table 900 to transform these data into frequency
- 22 components.
- 23 A frequency domain updating unit 410 first determines
- 24 whether additional information is embedded in the frequency

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- 1 components obtained by the MDCT/DFT transformer 230. If
- 2 additional information is embedded therein, the frequency
- 3 domain updating unit 410 further determines whether the
- 4 contents of the additional information should be changed.
- 5 Only when the contents of the additional information should
- 6 be changed is the updating of the additional information
- 7 performed for the frequency components (the determination
- 8 results may be output so that a user of the updating unit
- 9 410 can understand it).
- 10 In accordance with the window information extracted by the
- 11 MDCT coefficient recovery unit 210, a DFT/MDCT transformer
- 12 240 employs the table 900 to transform, into MDCT
- 13 coefficients sequence, the frequency components that have
- 14 been updated by the frequency domain updating unit 250.
- 15 Finally, an MDCT coefficient compressor 220 compresses the
- 16 MDCT coefficients sequence obtained by the DFT/MDCT
- 17 transformer 240, as well as the window information and the
- 18 other information that are extracted by the MDCT coefficient
- 19 recovery unit 210. The compressed audio data are thus
- 20 obtained. The prediction method, the inverse quantization
- 21 and the Huffmann decoding, which are designated in the
- 22 window and the other information, are employed for the data
- 23 compression.
- 24 General hardware arrangement
- 25 The apparatus and the systems according to the present

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- 1 invention can be carried out by using the hardware of a
- 2 common computer. Fig. 12 is a diagram illustrating the
- 3 hardware arrangement for a general personal computer. A
- 4 system 100 comprises a central processing unit (CPU) 1 and a
- 5 main memory 4. The CPU 1 and the main memory 4 communicate,
- 6 via a bus 2 and an IDE controller 25, with a hard disk drive
- 7 (HDD) 13, which is an auxiliary storage device (or a storage
- 8 medium drive, such as a CD-ROM 26 or a DVD 32). Similarly,
- 9 the CPU 1 and the main memory 4 communicate, via a bus 2 and
- 10 a SCSI controller 27, with a hard disk drive 30, which is an
- 11 auxiliary storage device (or a storage medium drive, such as
- 12 an MO 28, a CD-ROM 29 or a DVD 31). A floppy disk drive
- 13 (FDD) 20 (or an MO or a CD-ROM drive) is connected to the
- 14 bus 2 via a floppy disk controller (FDC) 19.
- 15 A floppy disk is inserted into the floppy disk drive 20.
- 16 Stored on the floppy disk and the hard disk drive 13 (or the
- 17 CD-ROM 26 or the DVD 32) are a computer program, a web
- 18 browser, the code for an operating system and other data
- 19 supplied in order that instructions can be issued to the CPU
- 20 1, in cooperation with the operating system and in order to
- 21 implement the present invention. These programs, code and
- 22 data are loaded into the main memory 4 for execution. The
- 23 computer program code can be compressed, or it can be
- 24 divided into a plurality of codes and recorded using a
- 25 plurality of media. The programs can also be stored on
- 26 another a storage medium, such as a disk, and the disk can
- 27 be driven by another computer.

- 1 The system 100 further includes user interface hardware.
- 2 User interface hardware components are, for example, a
- 3 pointing device (a mouse, a joy stick, etc.) 7 or a keyboard
- 4 6 for inputting data, and a display (CRT) 12. A printer,
- 5 via a parallel port 16, and a modem, via a serial port 15,
- 6 can be connected to the communication terminal 100, so that
- 7 it can communicate with another computer via the serial port
- 8 15 and the modem, or via a communication adaptor 18 (an
- 9 ethernet or a token ring card). A remote transceiver may be
- 10 connected to the serial port 15 or the parallel port 16 to
- 11 exchange data using ultraviolet rays or radio.
- 12 A loudspeaker 23 receives, through an amplifier 22, sounds
- 13 and tone signals that are obtained through D/A
- 14 (digital-analog) conversion performed by an audio controller
- 15 21, and releases them as sound or speech. The audio
- 16 controller 21 performs A/D (analog/digital) conversion for
- 17 sound information received via a microphone 24, and
- 18 transmits the external sound information to the system. The
- 19 sound may be input at the microphone 24, and the compressed
- 20 data produced by this invention may be generated based on
- 21 the sound that is input.
- 22 It would therefore be easily understood that the present
- 23 invention can be provided by employing an ordinary personal
- 24 computer (PC), a work station, a notebook PC, a palmtop PC,
- 25 a network computer, various types of electric home
- 26 appliances, such as a computer-incorporating television, a
- 27 game machine that includes a communication function, a

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- 1 telephone, a facsimile machine, a portable telephone, a PHS,
- 2 a PDA, another communication terminal, or a combination of
- 3 these apparatuses. The above described components, however,
- 4 are merely examples, and not all of them are required for
- 5 the present invention.
- 6 Advantages of the Invention
- 7 According to the present invention, provided is a method and
- 8 a system for embedding, detecting or updating additional
- 9 information embedded in compressed audio data, without
- 10 having to decompress the audio data. Further, according to
- 11 the method of the invention, the additional information
- 12 embedded in the compressed audio data can be detected using
- 13 a conventional watermarking technique, even when the audio
- 14 data have been decompressed.
- 15 The present invention can be realized in hardware, software,
- 16 or a combination of hardware and software. The present
- 17 invention can be realized in a centralized fashion in one
- 18 computer system, or in a distributed fashion where different
- 19 elements are spread across several interconnected computer
- 20 systems. Any kind of computer system or other apparatus
- 21 adapted for carrying out the methods described herein is
- 22 suitable. A typical combination of hardware and software
- 23 could be a general purpose computer system with a computer
- 24 program that, when being loaded and executed, controls the
- 25 computer system such that it carries out the methods
- 26 described herein. The present invention can also be embedded

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- 1 in a computer program product, which comprises all the
- 2 features enabling the implementation of the methods described
- 3 herein, and which when loaded in a computer system is
- 4 able to carry out these methods.
- 5 Computer program means or computer program in the present
- 6 context mean any expression, in any language, code or
- 7 notation, of a set of instructions intended to cause a system
- 8 having an information processing capability to perform a
- 9 particular function either directly or after conversion to
- 10 another language, code or notation and/or reproduction in a
- 11 different material form.
- 12 It is noted that the foregoing has outlined some of the more
- 13 pertinent objects and embodiments of the present invention.
- 14 This invention may be used for many applications. Thus,
- 15 although the description is made for particular arrangements
- 16 and methods, the intent and concept of the invention is
- 17 suitable and applicable to other arrangements and
- 18 applications. It will be clear to those skilled in the art
- 19 that other modifications to the disclosed embodiments can be
- 20 effected without departing from the spirit and scope of the
- 21 invention. The described embodiments ought to be construed
- 22 to be merely illustrative of some of the more prominent
- 23 features and applications of the invention. Other beneficial
- 24 results can be realized by applying the disclosed invention
- 25 in a different manner or modifying the invention in ways
- 26 known to those familiar with the art.

27

[Equation 1]
$$M_{k} = \sum_{n=0}^{N-1} X_{n} \cos \left\{ \frac{2\pi}{N} \left(n + \frac{N}{4} + \frac{1}{2} \right) \left(k + \frac{1}{2} \right) \right\}$$

['Equation 2]
$$R_{k} = \sum_{0}^{N-1} X_{n} \cos \left\{ \frac{2\pi}{N} \ln \right\}$$

$$I_{k} = -\sum_{0}^{N-1} X_{n} \sin \left\{ \frac{2\pi}{N} \ln \right\}$$

[Equation 3]
$$R = \sum_{b=1}^{n+1} T_{r,b}^{T} M_{b}$$

$$I = \sum_{b=1}^{n+1} T_{r,b}^{T} M_{b}$$

[Equation 4]
$$R = \sum_{b=1}^{n+1} T_{W,r,b}^{T} M_b$$

$$I = \sum_{b=1}^{n+1} T_{W,r,b}^{T} M_b$$

[Equation 5]
$$M_{b} = T_{r,b}R + T_{i,b}I$$

$$= \sum_{k=1}^{N/2} (R_{k}V_{r,b,k} + I_{k}V_{i,b,k})$$

[| Equation 6]
$$R_{k} = V_{r,b,k} \cdot M_{b}$$

$$I_{k} = V_{i,b,k} \cdot M_{b}$$

[a]
$$u_{r,b+m,k} = u_{r,b,k}$$

[b] $u_{r,b+m,k} = -u_{r,b,k}$
[c] $u_{r,b+m,k} = -u_{i,b,k}$
[d] $u_{r,b+m,k} = u_{i,b,k}$